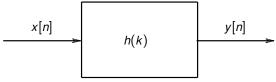
EDISP (Filters intro) (English) Digital Signal Processing Digital/Discrete Time/ filters introduction lecture

January 10, 2011

DT system properties

An LTI system is described with its impulse response



$$y(n) = \sum_{k=-\infty}^{\infty} h(k) \cdot x(n-k)$$

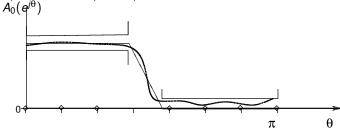
but we are interested in its characteristics in the frequency domain (frequency response)

$$H(e^{i\theta}) = H(z)|_{z=e^{i\theta}} = \sum_{n=-\infty}^{\infty} h(n)e^{-jn\theta}$$

$$\begin{array}{llll} \text{Magnitude of fr. response} & \textit{A}(\theta) & = & |\textit{H}(e^{i\theta})| \\ \text{Phase of fr. response} & \phi(\theta) & = & \textit{arg}[\textit{H}(e^{i\theta})] \\ \text{Group delay} & \tau(\theta) & = & -\textit{d}\phi(\theta)/\textit{d}\theta \end{array}$$

Filter design

Specification: stopband, passband, tolerances



Approximation: find best rational function

$$\frac{b_0 + b_1 z^{-1} + \ldots + b_M z^{-M}}{1 + a_1 z^{-1} + \ldots + a_N z^{-N}}$$
 (IIR)

or

$$b_0 + b_1 z^{-1} + \ldots + b_M z^{-M}$$
 (FIR)

determine order and coefficients, check stability

Implementation: structure, noise, hardware/software . . .

FIR filter design - window method

- $\qquad \qquad \text{Ideal filter: } A_0(\theta) = \left\{ \begin{array}{ll} 1 & \text{for} & |\theta| < \theta_p \\ 0 & \text{for} & \theta_p < |\theta| \leq \pi \end{array} \right. \text{ and zero phase}$
- Impulse response:

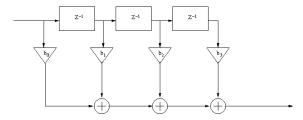
$$h_0(n) = \frac{1}{2\pi} \, \int_{-\pi}^{\pi} \, H_0\left(e^{i\theta}\,\right) \, e^{jn\theta} \, d\theta \, = \, \frac{\theta_p}{\pi} \, \frac{\sin n\theta_p}{n\,\theta_p}$$

is non-causal and infinite!

- ▶ Make it finite: $h_P[n] = h_0[n]g[n](g[n] = 0 \text{ for } |n| > P)$
- ▶ Shift it to be causal delay by P samples: $h[n] = h_P[n-P]$
- finally we obtain

$$H(z) = \sum_{n=0}^{2P} h(n) z^{-n} = z^{-P} H_P(z)$$

Implementations

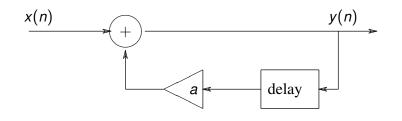


- Transversal (linear convolution)
- ▶ By IFFT(FFT(x) · FFT(h))
- ▶ Lagrange structure: FIR with IIR inside

$$H(z) = \frac{(1 - z^{-N})}{N} \sum_{k=0}^{N-1} \frac{H(k)}{1 - e^{-j2\pi k/N} z^{-1}}$$

where H(k) are samples of frequency response at $\theta_k = 2\pi K/N$; efficient if many H(k) are almost zero.

Example



$$y(n) = x(n) + ay(n-1)$$

$$Y(z) = X(z) + aY(z)z^{-1}$$

$$Y(z) = X(z)\frac{1}{1 - az^{-1}}$$

$$H(z) = \frac{Y(z)}{X(z)} = \frac{1}{1 - az^{-1}} = \frac{z}{z - a}$$

so $h(n) = a^n u(n)$ if we assume causal solution.

Example cont.

$$H(z) = \frac{z}{z - a}$$

$$H(e^{i\theta}) = \frac{e^{i\theta}}{e^{i\theta} - a}$$

$$|H(e^{i\theta})| = \frac{1}{|e^{i\theta} - a|}$$

$$|H(e^{i\theta})| = \frac{1}{|\cos\theta - a + j\sin\theta|} = \frac{1}{\sqrt{1 - 2a \cdot \cos\theta + a^2}}$$

